

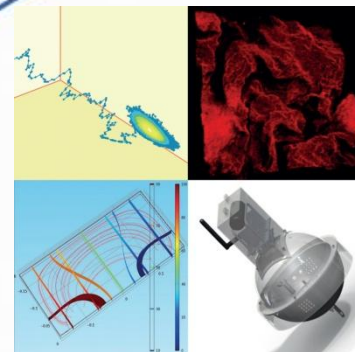


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# Active induction balance method for metal detector sensing head utilizing transmitter-bucking and dual current source

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Sensors & their Applications XVII

# Outline

- Introduction
- Induction balance problem
- Sensing head – design and modeling
- Active induction balance technique
- Experiments and results
- Conclusions



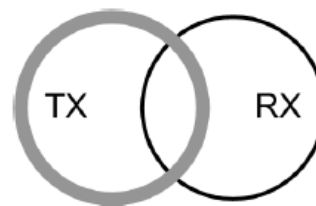
# Background and motivation

- Electromagnetic induction (EMI) sensors in humanitarian demining → still an area of active research!
- Time-domain (TD) EMI sensors:
  - Inherently balanced, but excitation spectrum limited!
- Frequency-domain (FD) EMI sensors:
  - Higher sensitivity and improved SNR,
  - Induction balance (IB) problem (i.e. direct inductive coupling between TX and RX coil) needs to be solved.

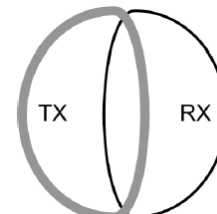


# Induction balance problem

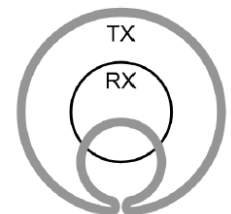
- Suppression of primary (excitation) field achieved by sensing head geometry:
- Physical separation of coils,
- Gradiometer configuration of RX coils,
- Overlapping coils,
- Orthogonal coils,
- Transmitter-bucking,
- ...



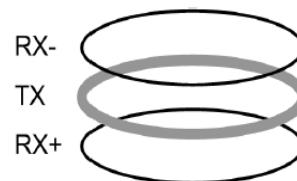
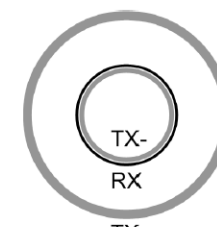
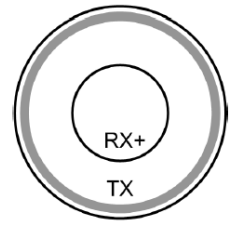
a) "OO" coils



b) "DD" coils

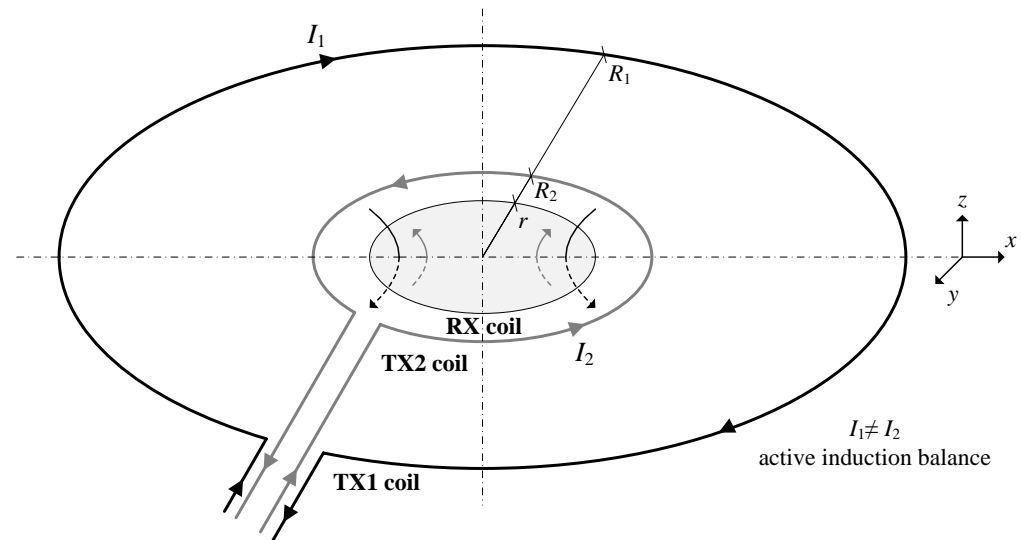


c) "4B" coils

d) coaxial coils,  
gradiometer conf. RXe) concentric coils,  
gradiometer conf. TXf) concentric coils,  
gradiometer conf. RX

# Sensing head - design objectives

- Ultimate objective: handheld FD EMI landmine detector featuring model-based metal characterization and soil compensation.
- Initial design requirements:
  - High sensitivity and dynamic range,
  - Simple and compact geometry,
  - High spatial resolution,
  - Pinpointing accuracy,
  - Good invertibility of measured data.



➔ **transmitter-bucking configuration**



# Sensing head model

- Vertical component of magnetic field  $B_z$  as a function of radial distance  $r$  from the coil centre:

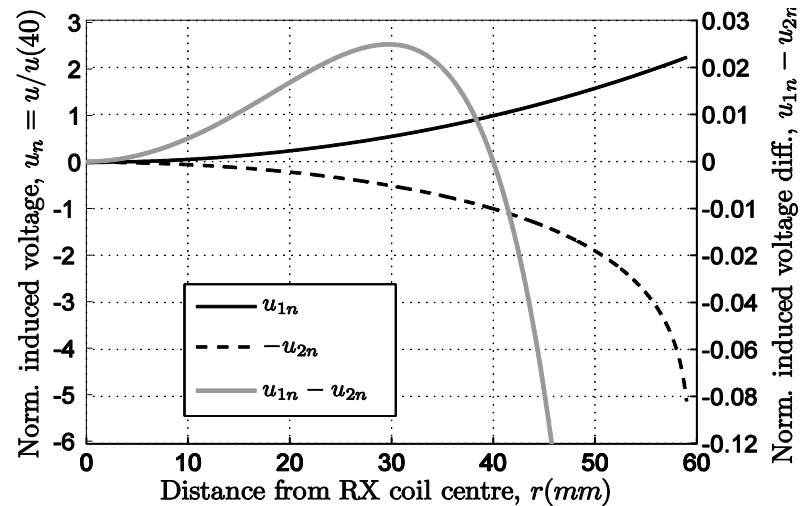
$$B_z(r) = \frac{\mu_0 I N}{2R} \left[ 1 + \sum_{n=1}^{\infty} \left[ \frac{(2n-1)!!}{(2n)!!} \right]^2 (2n+1) \left( \frac{r}{R} \right)^{2n} \right]$$

for  $r \leq R$ , inside loop (circular coil approx.)

- For a detector coil of radius  $r$  IB is obtained if:

$$\int_0^r B_z^1(r) 2\pi r dr = \int_0^r B_z^2(r) 2\pi r dr$$

$$\frac{N_1}{R_1} \sum_{n=0}^{\infty} \left[ \frac{(2n-1)!!}{(2n)!!} \right]^2 \frac{(2n+1)}{(2n+2)} \left( \frac{r}{R_1} \right)^{2n} = \frac{N_2}{R_2} \sum_{n=0}^{\infty} \left[ \frac{(2n-1)!!}{(2n)!!} \right]^2 \frac{(2n+1)}{(2n+2)} \left( \frac{r}{R_2} \right)^{2n}$$



Normalized voltages induced in RX coil in response to TX1 and TX2 coils.



# Induction balance sensitivity analysis

- IB sensitivity to small perturbations of geometrical properties of coils ( $R_1, R_2, r$ ) ?

$$S_x^u = \frac{\Delta u}{u_{IB}} \left( \frac{\Delta x}{x} \right)^{-1} = \left( \frac{\Delta u_1}{u_1} - \frac{\Delta u_2}{u_2} \right) \left( \frac{\Delta x}{x} \right)^{-1} = S_x^{u_1} - S_x^{u_2}$$

- For a given geometry:  $S_{R_1}^u = -1.057$ ,  $S_{R_2}^u = -1.555$ ,  $S_r^u = 0.498$

TX1 coil radius		TX2 coil radius		RX coil radius		Induced voltage (excitation current, I=1A)		
							f=1kHz	f=100kHz
$\Delta R_1/R_1$ (%)	$\Delta R_1$ (mm)	$\Delta R_2/R_2$ (%)	$\Delta R_2$ (mm)	$\Delta r/r$ (%)	$\Delta r$ (mm)	$\Delta u/u_{IB}$ (%)	$\Delta u$ , (mV)	$\Delta u$ , (V)
1	1.5	0	0.0	0	0.0	-1.057	3.1	0.310
2	3.0	0	0.0	0	0.0	-2.114	6.2	0.621
3	4.5	0	0.0	0	0.0	-3.171	9.3	0.931
0	0.0	1	0.6	0	0.0	-1.555	4.6	0.457
0	0.0	2	1.2	0	0.0	-3.110	9.1	0.913
0	0.0	3	1.8	0	0.0	-7.775	22.8	2.284
0	0.0	0	0.0	1	0.4	0.498	1.5	0.146
0	0.0	0	0.0	2	0.8	0.996	2.9	0.293
0	0.0	0	0.0	5	2.0	2.490	7.3	0.731



# Active induction balance (AIB)

- Excitation current in each transmitter coil controlled separately:

$$\frac{I_1 N_1}{R_1} \sum_{n=0}^{\infty} \left[ \frac{(2n-1)!!}{(2n)!!} \right]^2 \frac{(2n+1)}{(2n+2)} \left( \frac{r}{R_1} \right)^{2n} = \frac{I_2 N_2}{R_2} \sum_{n=0}^{\infty} \left[ \frac{(2n-1)!!}{(2n)!!} \right]^2 \frac{(2n+1)}{(2n+2)} \left( \frac{r}{R_2} \right)^{2n}$$

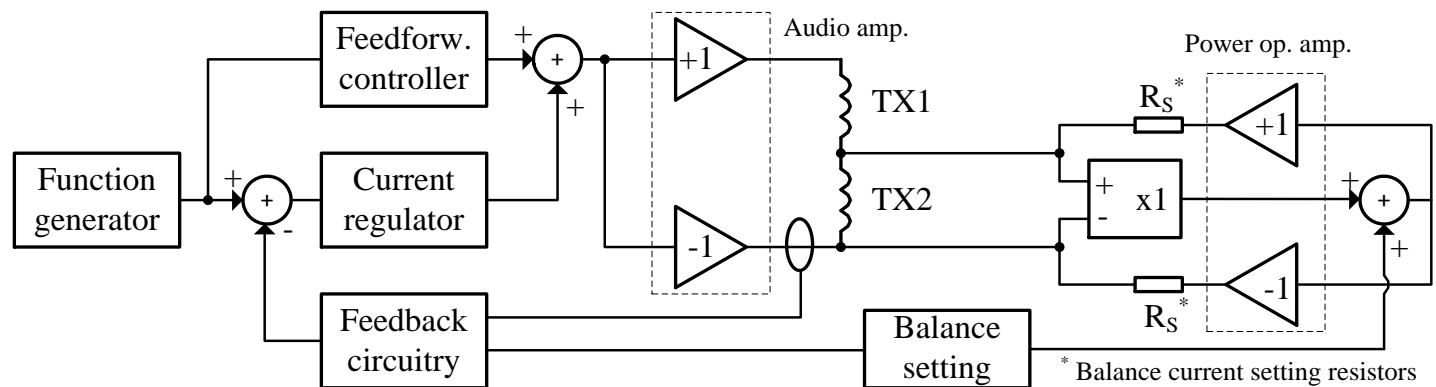
- Motivation:
  - Compensation of small imperfections of sensing head geometry and the effects of finite size coils,
  - Sensing head easier to produce,
  - Potential for more efficient soil compensation (lower loss of detector sensitivity / dynamic range).





# AIB implementation

- Transmitter coil driven by current source → transmitted field unaffected by changes in coil impedance, soil properties, lift-off and orientation of the sensing head.
- Dual current source scheme:
  - Main (excitation) current source drives both TX coils,
  - Balancing current source additionally drives only inner TX coil.

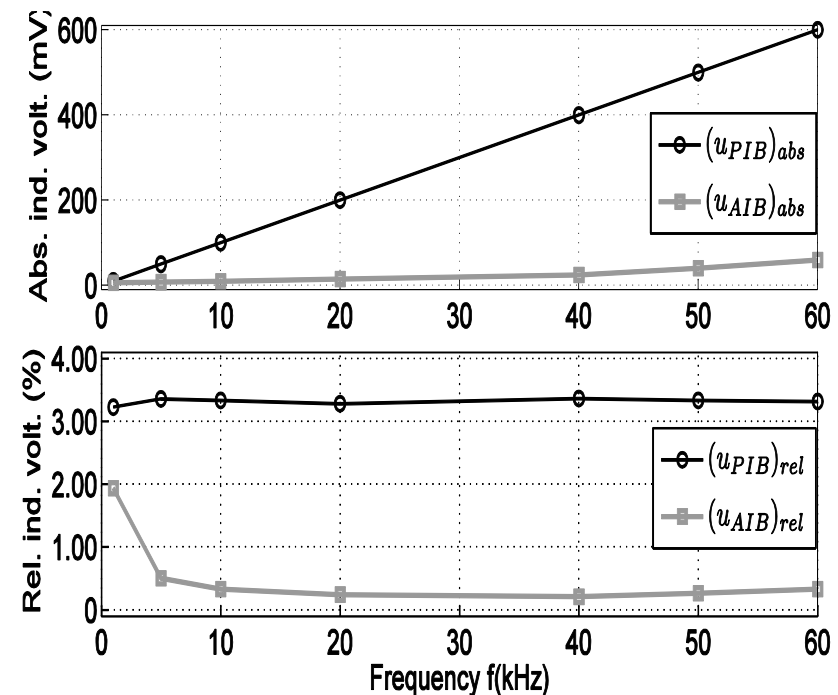
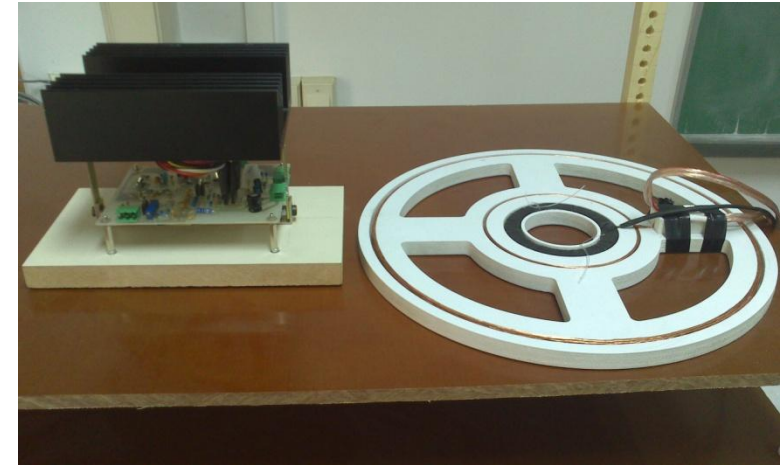


\* Balance current setting resistors



# Experiments and results

- Dual current source and sensing head in transmitter-bucking configuration implemented as laboratory prototypes.
- Induced voltage imbalance measured for passive IB and AIB at different frequencies (in absolute and relative terms).
- Residual imbalances from passive IB can be effectively compensated by AIB.



# Conclusions

- For a design of novel, frequency-domain EMI landmine detector, we propose a sensing head configuration based on the transmitter-bucking approach.
- Overall, IB sensitivities to small perturbations of sensing head geometrical properties are rather low.
- Total sensor imbalances in absolute terms can become large, resulting in significant loss of sensitivity / dynamic range.
- Prototype sensor with AIB and dual current source overcomes the limitations of passive IB.
- Future work: further characterisation of the method, automatic compensation of soil-related imbalances.





THANK YOU FOR YOUR ATTENTION!

